Lecture – 10

Conversion Between State Space and Transfer Function Representations in Linear Systems – II

Dr. Radhakant Padhi

Asst. Professor

Dept. of Aerospace Engineering Indian Institute of Science - Bangalore



$$H(s) = \left[\frac{b_n s^n + b_{n-1} s^{n-1} + \dots + b_1 s + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0} \right] = \frac{y(s)}{u(s)}$$

i.e.

$$y^{(n)} + a_{n-1}y^{(n-1)} + \dots + a_1\dot{y} + a_0y$$

$$= b_0u + b_1\dot{u} + \dots + b_{n-1}u^{(n-1)} + b_nu^{(n)}$$

Ref: B. Friedland, Control System Design Mc Graw Hill, 1986.

Define the state variables x_1, \dots, x_n such that

$$y = x_1 + p_0 u$$

$$\dot{x}_1 = x_2 + p_1 u$$

$$\dot{x}_2 = x_3 + p_2 u$$

$$\vdots$$

$$\dot{x}_{n-1} = x_n + p_{n-1} u$$

$$\dot{x}_n = -a_{n-1} x_n - \dots - a_0 x_1 + p_n u$$

From the above definition, we have

$$y = x_{1} + p_{0}u$$

$$\dot{y} = \dot{x}_{1} + p_{0}\dot{u} = (x_{2} + p_{1}u) + p_{0}\dot{u}$$

$$\ddot{y} = \dot{x}_{2} + p_{1}\dot{u} + p_{0}\ddot{u} = (x_{3} + p_{2}u) + p_{1}\dot{u} + p_{0}\ddot{u}$$

$$\vdots$$

$$y^{(n-1)} = x_{n} + p_{n-1}u + p_{n-2}\dot{u} + \dots + p_{1}u^{(n-2)} + p_{0}u^{(n-1)}$$

$$y^{(n)} = -a_{n-1}x_{n} - a_{n-2}x_{n-1} - \dots - a_{0}x_{1} + p_{n}u$$

$$+ p_{n-1}\dot{u} + p_{n-2}\ddot{u} + \dots + p_{1}u^{(n-1)} + p_{0}u^{(n)}$$

$$y^{(n)} + a_{n-1}y^{(n-1)} + \dots + a_{1}\dot{y} + a_{0}y$$

$$= (p_{n} + a_{n-1}p_{n-1} + \dots + a_{1}p_{1} + a_{0}p_{0})u$$

$$+ (p_{n-1} + \dots + a_{2}p_{1} + a_{1}p_{0})\dot{u}$$

$$+ \dots$$

$$+ (p_{1} + a_{n-1}p_{0})u^{(n-1)}$$

$$+ p_{0}u^{(n)}$$

$$= b_{0}u + b_{1}\dot{u} + \dots + b_{n-1}u^{(n-1)} + b_{n}u^{(n)}$$
(from the TF)

Equating the coefficients:

$$p_0 = b_n$$

$$p_1 + a_{n-1}p_0 = b_{n-1}$$

$$\vdots$$

$$p_{n-1} + a_{n-1}p_{n-1} + \dots + a_1p_0 = b_1$$
$$p_n + a_{n-1}p_{n-1} + \dots + a_0p_0 = b_0$$

Now, we need to solve for $(p_0, p_1, ..., p_n)$

In matrix form:

$$\begin{bmatrix} 1 & 0 & 0 & \cdots & 0 \\ a_{n-1} & 1 & 0 & \cdots & 0 \\ a_{n-2} & a_{n-1} & \ddots & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ a_0 & a_1 & \cdots & a_{n-1} & 1 \end{bmatrix} \begin{bmatrix} p_0 \\ p_1 \\ \vdots \\ \vdots \\ p_n \end{bmatrix} = \begin{bmatrix} b_n \\ b_{n-1} \\ \vdots \\ \vdots \\ b_0 \end{bmatrix}$$

Toeplitz Matrix

Note: Toeplitz matrix is a nonsinular matrix $(determinant = 1 \neq 0)$

Hence, the solution always exists!

State space form (from definition of state & output variables):

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -a_0 & -a_1 & \cdots & -a_{n-1} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} p_0 \\ p_1 \\ \vdots \\ p_n \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 0 & \cdots & 0 \end{bmatrix} X + \begin{bmatrix} p_0 \end{bmatrix} u$$

Note: The physical meaning of state variables in different companion froms are different.

Block Diagram for realization of Toeplitz first companion form

Alternate/Toeplitz first companion form: Some comments

- Toeplitz realization also requires 'n' integrators
- Extension of Toeplitz realization is straightforward to MISO systems
 - One need to solve m-system of linear equations, where m is the number of input (control) variables.
 - However, only n integrators are needed.

(Observable Canonical Form)

$$H(s) = \left[\frac{b_n s^n + b_{n-1} s^{n-1} + \dots + b_1 s + b_0}{s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0} \right] = \frac{y(s)}{u(s)}$$

i.e.

$$\left(s^{n} + a_{n-1}s^{n-1} + \dots + a_{1}s + a_{0} \right) y(s)$$

$$= \left(b_{n}s^{n} + b_{n-1}s^{n-1} + \dots + b_{1}s + b_{0} \right) u(s)$$

Rearranging the terms:

$$s^{n} [y(s) - b_{n}u(s)] + s^{n-1} [a_{n-1}y(s) - b_{n-1}u(s)] + \cdots$$
$$\cdots + [a_{0}y(s) - b_{0}u(s)] = 0$$

(Observable Canonical Form)

Simplify:

$$[y(s)-b_n u(s)] = \frac{1}{s} [b_{n-1} u(s) - a_{n-1} y(s)] + \dots + \frac{1}{s^n} [b_0 u(s) - a_0 y(s)]$$

Solve for y(s):

$$y(s) = b_n u(s) + \frac{1}{s} [b_{n-1} u(s) - a_{n-1} y(s)] + \dots + \frac{1}{s^n} [b_0 u(s) - a_0 y(s)]$$

$$= b_n u(s) + \frac{1}{s} [b_{n-1} u(s) - a_{n-1} y(s)] + \underbrace{\frac{1}{s} [[b_{n-2} u(s) - a_{n-2} y(s)] \dots +]}_{s} \dots$$

Block Diagram Realization

(Observable Canonical Form)

The equations can be written as:

$$y = x_1 + b_n u$$

$$\dot{x}_1 = (x_2 - a_{n-1}y + b_{n-1}u) = -a_{n-1}x_1 + x_2 + (b_{n-1} - a_{n-1}b_n)u$$

$$\dot{x}_2 = (x_3 - a_{n-2}y + b_{n-2}u) = -a_{n-2}x_1 + x_3 + (b_{n-2} - a_{n-2}b_n)u$$

$$\vdots$$

$$\dot{x}_{n-1} = (x_n - a_1y + b_1u) = -a_1x_1 + x_n + (b_1 - a_1b_n)u$$

$$\dot{x}_n = b_0u - a_0y = -a_0x_1 + (b_0 - a_0b_n)u$$

(Observable Canonical Form)

In vector-matrix form, we can write it as

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_{n-1} \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} -a_{n-1} & 1 & 0 & 0 & \cdots & 0 \\ -a_{n-2} & 0 & 1 & 0 & \cdots & 0 \\ \vdots & \vdots & \vdots & \vdots & \ddots & \cdots & \vdots \\ -a_1 & 0 & 0 & 0 & \cdots & 1 \\ -a_0 & 0 & 0 & 0 & \cdots & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} + \begin{bmatrix} b_{n-1} - a_{n-1}b_n \\ b_{n-2} - a_{n-2}b_n \\ \vdots \\ b_1 - a_1b_n \\ b_0 - a_0b_n \end{bmatrix} u$$

$$y = \underbrace{\begin{bmatrix} 1 & 0 & \cdots & 0 \end{bmatrix}}_{C} X + \underbrace{\begin{bmatrix} b_n \end{bmatrix}}_{D} u$$

Second Companion Form (Observable Canonical Form)

- (1) In second companion form, the coefficients of the denominator of the transfer function appear in a column, whereas in the first companion form they appear in a row.
- (2) Extensions of second companion form to SIMO,MISO cases etc. are possible (but beyond the scope this course)
- (3) Controllable canonical form is good for control design, whereas observable canonical form is good for observer/filter design

Comment on minimal realization (for MIMO systems)

We have seen realization of:

> SIMO systems: *n* integrators

> MISO systems: *n* integrators

Question: How about MIMO systems? Will it still be possible to realize it with *n* integrators?

Answer: No!

However, one way of realizing MIMO systems will be to use a number of structures (of either SIMO or MISO form) in parallel; *i.e.* if $U \in \mathbb{R}^m$ and $Y \in \mathbb{R}^p$

then it is always possible to realize such a MIMO system with no more than *n x min (m, p)* integrators.

Comment on minimal realization (for MIMO systems)

Question: How about fewer integrators?

Answer: This leads to the questions of "<u>minimal realization</u>"; a subject of considerable research during 1970s.

Why necessary? Because a non-minimal realization is either <u>non-controllable</u> or <u>non-observable</u> (or both). It may cause theoretical and computational problems too.

Solution: Several algorithms exist for finding a minimal realization and corresponding *A,B,C,D* matrices. However, these are beyond the scope of this course!

Further reading: T. Kailath, Linear Systems, Prentice Hall, 1980.

